

# Lecture - 6

## 1.3 Properties of DFT :

In this section we will study some important properties of DFT. We know that, the DFT of discrete time sequence,  $x(n)$  is denoted by  $X(k)$ . And the DFT and IDFT pair is denoted by,

$$\begin{array}{c} \text{DFT} \\ x(n) \longleftrightarrow X(k) \\ \text{N} \end{array}$$

Here 'N' indicates 'N' point DFT.

### 1.3.1 Linearity :

**Statement :** If  $\begin{array}{c} \text{DFT} \\ x_1(n) \longleftrightarrow X_1(k) \\ \text{N} \end{array}$  and  $\begin{array}{c} \text{DFT} \\ x_2(n) \longleftrightarrow X_2(k) \\ \text{N} \end{array}$  then,

$$\begin{array}{c} \text{DFT} \\ a_1 x_1(n) + a_2 x_2(n) \longleftrightarrow a_1 X_1(k) + a_2 X_2(k) \\ \text{N} \end{array}$$

Here  $a_1$  and  $a_2$  are some constants.

**Proof :** According to the definition of DFT,

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad \dots(1)$$

Here  $x(n) = a_1 x_1(n) + a_2 x_2(n)$

$$\begin{aligned} \therefore X(k) &= \sum_{n=0}^{N-1} [a_1 x_1(n) + a_2 x_2(n)] W_N^{kn} \\ &= \sum_{n=0}^{N-1} a_1 x_1(n) W_N^{kn} + \sum_{n=0}^{N-1} a_2 x_2(n) W_N^{kn} \end{aligned}$$

Since  $a_1$  and  $a_2$  are constants; we can take it out of the summation sign.

$$\therefore X(k) = a_1 \sum_{n=0}^{N-1} x_1(n) W_N^{kn} + a_2 \sum_{n=0}^{N-1} x_2(n) W_N^{kn} \quad \dots(2)$$

Comparing Equation (2) with the definition of DFT,

$$X(k) = a_1 X_1(k) + a_2 X_2(k)$$

**Meaning :** DFT of linear combination of two or more signals is equal to the sum of linear combination of DFT of individual signal.

### 1.3.2 Periodicity :

**Statement :** If  $x(n) \xleftrightarrow[N]{\text{DFT}} X(k)$  then

$$x(n+N) = x(n) \quad \text{for all } n$$

$$\text{and } X(k+N) = X(k) \quad \text{for all } k.$$

**Proof :** According to the definition of DFT,

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad \dots(1)$$

Replacing  $k$  by  $k+N$  we get,

$$X(k+N) = \sum_{n=0}^{N-1} x(n) W_N^{(k+N)n}$$

$$\therefore X(k+N) = \sum_{n=0}^{N-1} x(n) W_N^{kn} W_N^{Nn} \quad \dots(2)$$

We know that  $W_N$  is a twiddle factor and it is given by,

$$W_N = e^{-j \frac{2\pi}{N}}$$

$$\therefore W_N^{Nn} = \left( e^{-j \frac{2\pi}{N}} \right)^{Nn} = e^{-j \frac{2\pi}{N} \cdot Nn} = e^{-j 2\pi n}$$

$$\therefore W_N^{Nn} = \cos 2\pi n - j \sin 2\pi n \quad \dots(3)$$

But 'n' is an integer  $\therefore \cos 2\pi n = 1$  and  $\sin 2\pi n = 0$

$$\therefore W_N^{Nn} = 1 \quad \dots(4)$$

Putting this value in Equation (2)

$$X(k+N) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad \dots(5)$$

Comparing Equations (1) and (5),

$$X(k+N) = X(k)$$

Hence proved

**Meaning :** DFT of a finite length sequence results in a periodic sequence.

### 1.3.3 Circular Symmetries of a Sequence :

We have studied the periodicity property of DFT. Suppose input discrete time sequence is  $x(n)$  then, the periodic sequence is denoted by  $x_p(n)$ . The period of  $x_p(n)$  is 'N' which means after 'N' the sequence  $x(n)$  repeats itself. Now we can write the sequence  $x_p(n)$  as,

$$x_p(n) = \sum_{l=-\infty}^{\infty} x(n-lN) \quad \dots(1)$$

We will consider one example. Let  $x(n) = \{1, 2, 3, 4\}$ . This sequence is shown in Fig. F-8(a). The periodic sequence  $x_p(n)$  is shown in Fig. F-8(b).

We will delay the periodic sequence  $x_p(n)$  by two sample as shown in Fig. F-8(c). This sequence is denoted by  $x_p(n-2)$ . Now the original signal is present in the range  $n = 0$  to  $n = 3$ . In the same range we will write the shifted signal as shown in Fig. F-8(d). This signal is denoted by  $x'(n)$ .

Now from Fig. F-8 we can write every sequence as follows :

$$x(n) = \{1, 2, 3, 4\} \quad \dots(2)$$

$$x_p(n) = \{ \dots 1, 2, 3, 4, 1, 2, 3, 4, 1, 2, 3, 4, \dots \} \quad \dots(3)$$

$$x_p(n-2) = \{ \dots 1, 2, 3, 4, 1, 2, 3, 4, 1, 2, 3, 4, \dots \} \quad \dots(4)$$

$$\text{and } x'(n) = \{3, 4, 1, 2\} \quad \dots(5)$$

Now observe Equations (2) and (5). We can say that the sequence  $x'(n)$  is obtained by circularly shifting sequence  $x(n)$ , by two samples. That means  $x'(n)$  is related to  $x(n)$  by circular shift.

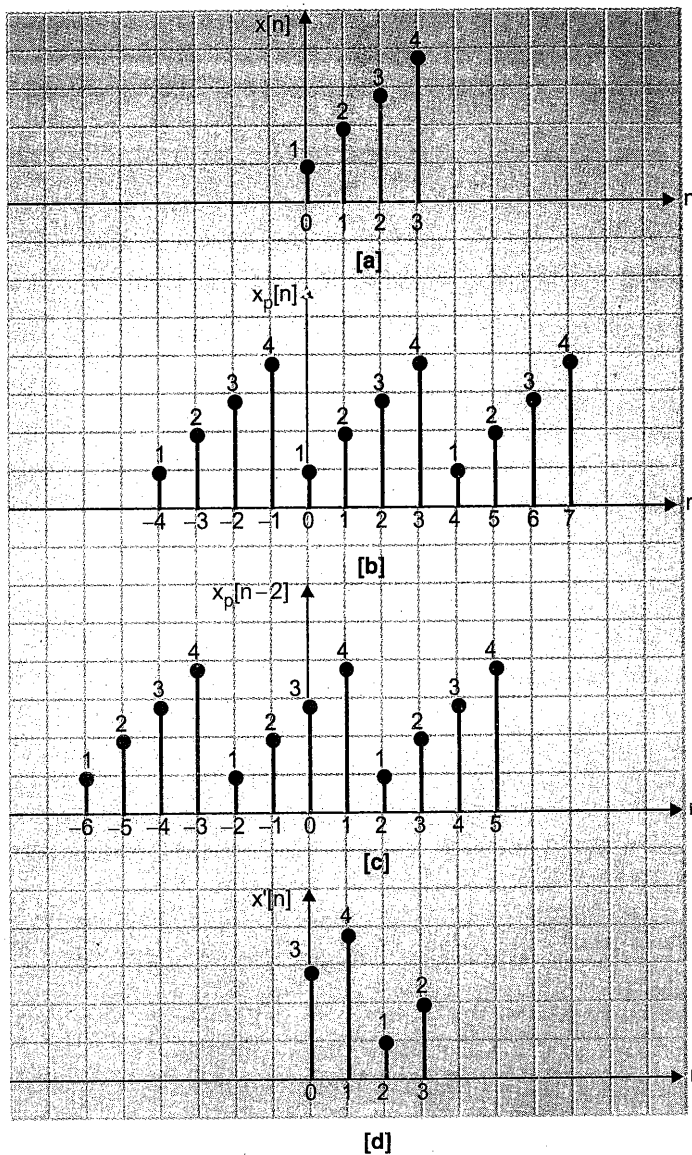


Fig. F-8 : Shifting of sequence  $x(n)$

**Notation :**

This relation of circular shift is denoted by,

$$x'(n) = x(n-k, \text{modulo } N) \quad \dots(6)$$

It means that divide  $(n-k)$  by  $N$  and retain the remainder only. We can also use the short hand notation as follows :

$$x'(n) = x((n-k))_N \quad \dots(7)$$

Here 'k' indicates the number of samples by which  $x(n)$  is delayed and 'N' indicates N-point DFT. In the present example, the sequence  $x(n)$  is delayed by two samples; thus  $k = 2$ . Since there are four samples in  $x(n)$ ; this is four point DFT. Thus  $N = 4$ . Now for this example Equation (6) becomes,

$$x'(n) = x((n-2))_4 \quad \dots(8)$$

### Graphical Representation :

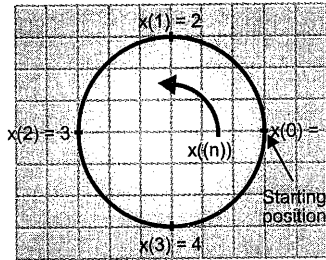
The circular shifting of a sequence is plotted graphically as follows :

(1) Circular plot of a sequence  $x(n)$  : Here we have considered,

$$x(n) = \{1, 2, 3, 4\}$$

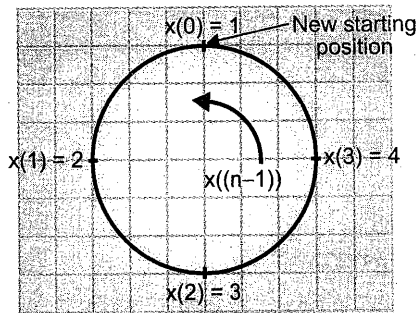
↑

Circular plot of  $x(n)$  is denoted by  $x((n))_4$ . This plot is obtained by writing the samples of  $x(n)$  circularly anticlockwise. It is shown in Fig. F-9(a).



**Fig. F-9(a) :  $x((n))_4$  – The samples of  $x(n)$  are plotted circularly anticlockwise**

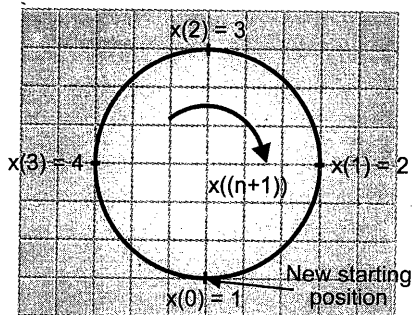
(2) Circular delay by one sample : To delay sequence  $x(n)$  circularly by one sample; shift every sample circularly in anticlockwise direction by 1. This is shown in Fig. F-9(b). This operation is denoted by  $x((n-1))$ .



**Fig. F-9(b) :  $x((n-1))$  shift every sample by 1 in anticlockwise direction**

Remember that delay by 'k' samples means shift the sequence circularly in anticlockwise direction by k.

(3) **Circular advance by one sample :** To advance sequence  $x(n)$  circularly by one sample; shift every sample circularly in clockwise direction by 1 sample. This sequence is denoted by  $x((n+1))$ . It is shown in Fig. F-9(c).



**Fig. F-9(c) :**  $x((n+1))$  shift every sample by one in clockwise direction

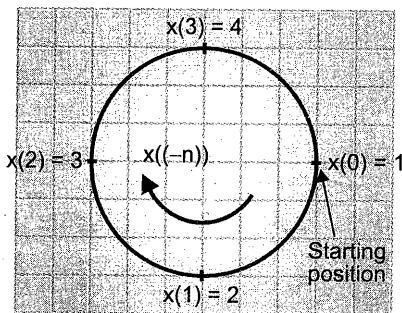
Remember that advance by  $k$  samples means shift the sequence circularly in clockwise direction by  $k$ .

(4) **Circularly folded sequence :** A circularly folded sequence is denoted by  $x((-n))$ . We have plotted sequence  $x((n))$  in anticlockwise direction. So folded sequence  $x((-n))$  is plotted in clockwise direction. It is shown in Fig. F-9(d).

Remember that circular folding means plot the samples in clockwise direction.

Now recall Equation (7) it is,

$$x'(n) = x((n-2))_4$$



**Fig. F-9(d) :**  $x((-n))$  samples are plotted circularly clockwise

It indicates delay of sequence  $x(n)$  by two samples. It is obtained by rotating samples of Fig. F-9(a) in anticlockwise direction by two samples. This sequence is shown in Fig. F-9(e):

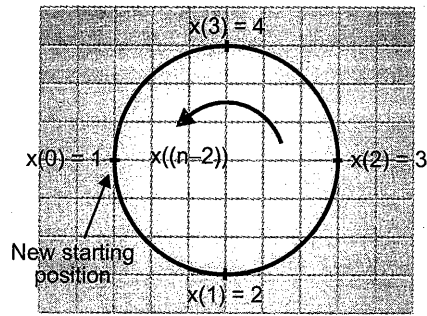


Fig. F-9(e) : Plot of  $x((n-2))_4$

(5) Circularly even sequence :

The  $N$ -point discrete time sequence is circularly even if it is symmetric about the point zero on the circle.

That means,

$$x(N-n) = x(n), \quad 1 \leq n \leq N-1$$

Consider the sequence,

$$x(n) = \{1, 4, 3, 4\}. \text{ It is plotted as shown in Fig. F-9(f).}$$

Note that this sequence is symmetric about point zero on the circle. So it is circularly even sequence. We can also verify it using mathematical equation,

The sequence is  $x(n) = \{1, 4, 3, 4\}$

$$\therefore x(0) = 1, \quad x(1) = 4, \quad x(2) = 3 \text{ and } x(3) = 4$$

We have the condition for circularly even sequence,

$$x(N-n) = x(n) \quad \dots(1)$$

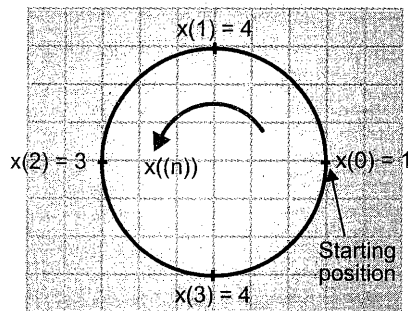


Fig. F-9(f) :  $x(n) = \{1, 4, 3, 4\}$

Here  $N = 4$ . We will check this condition by putting different values of  $n$  as follows :

$$\text{For } n = 1 \Rightarrow x(4-1) = x(1) \text{ that means } x(3) = x(1) = 4$$



For  $n = 2 \Rightarrow x(4-2) = x(2)$  that means  $x(2) = x(2) = 3$

For  $n = 3 \Rightarrow x(4-3) = x(3)$  that means  $x(1) = x(3) = 4$ .

Since for all values of 'n', Equation (1) is satisfied; the given sequence is circularly even.

**(6) Circularly odd sequence :**

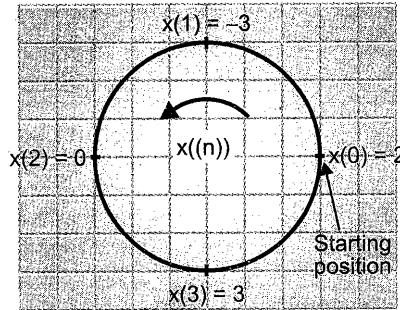
A N-point sequence is called circularly odd if it is antisymmetric about point zero on the circle.

That means,

$$x(N-n) = -x(n), \quad 1 \leq n \leq N-1$$

Consider the sequence,  $x(n) = \{2, -3, 0, 3\}$

This sequence is plotted as shown in Fig. F-9.(g)



**Fig. F-9(g) : Plot of  $x(n) = \{2, -3, 0, 3\}$**

Here  $x(0) = 2$ ,  $x(1) = -3$ ,  $x(2) = 0$  and  $x(3) = 3$ .

We have the condition for circularly odd sequence,

$$x(N-n) = -x(n), \quad \text{for } 1 \leq n \leq N-1 \quad \dots(1)$$

For  $n = 1 \Rightarrow x(4-1) = -x(1)$  that means  $x(3) = -x(1)$

For  $n = 2 \Rightarrow x(4-2) = -x(2)$  that means  $x(2) = -x(2)$

For  $n = 3 \Rightarrow x(4-3) = -x(3)$  that means  $x(1) = -x(3)$

Thus for all values of 'n', Equation (1) is satisfied. Hence the sequence is circularly odd.

### 1.3.7 Circular Convolution :

**Statement :** The multiplication of two DFTs is equivalent to the circular convolution of their sequences in time domain.

**Mathematical equation :**

$$\begin{aligned} \text{If } x_1(n) \xleftrightarrow[N]{\text{DFT}} X_1(k) \text{ and } x_2(n) \xleftrightarrow[N]{\text{DFT}} X_2(k) \text{ then,} \\ x_1(n) \circledast x_2(n) \xleftrightarrow[N]{\text{DFT}} X_1(k) \cdot X_2(k) \end{aligned} \quad \dots(1)$$

Here  $\circledast$  indicates circular convolution.

Let the result of circular convolution of  $x_1(n)$  and  $x_2(n)$  be  $y(m)$  then the circular convolution can also be expressed as,

$$y(m) = \sum_{n=0}^{N-1} x_1(n) x_2((m-n))_N, \quad m = 0, 1, \dots, N-1 \quad \dots(2)$$

Here the term  $x_2((m-n))_N$  indicates the circular convolution.

**Proof :** Consider two discrete time sequences  $x_1(n)$  and  $x_2(n)$ . The DFT of  $x_1(n)$  can be expressed as,

$$X_1(k) = \sum_{n=0}^{N-1} x_1(n) e^{-j \frac{2\pi kn}{N}}, \quad k = 0, 1, \dots, N-1 \quad \dots(3)$$

To avoid the confusion we will write the DFT of  $x_2(n)$  using different index of summation.

$$\therefore X_2(k) = \sum_{l=0}^{N-1} x_2(n) e^{-j \frac{2\pi kl}{N}}, \quad k = 0, 1, \dots, N-1 \quad \dots(4)$$

Note that in Equation (4); instead of 'n' we have used 'l'.

We will denote the multiplication of two DFTs;  $X_1(k)$  and  $X_2(k)$  by  $Y(k)$ .

$$\therefore Y(k) = X_1(k) \cdot X_2(k) \quad \dots(5)$$

Let IDFT of  $Y(k)$  be  $y(m)$ . Then using definition of IDFT,

$$y(m) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) e^{j \frac{2\pi k m}{N}} \quad \dots(6)$$

Putting Equation (5) in Equation (6),

$$y(m) = \frac{1}{N} \sum_{k=0}^{N-1} X_1(k) \cdot X_2(k) e^{j \frac{2\pi k m}{N}} \quad \dots(7)$$

Putting the values of  $X_1(k)$  and  $X_2(k)$  from Equations (3) and (4) in Equation (7) we get,

$$y(m) = \frac{1}{N} \sum_{k=0}^{N-1} \left[ \sum_{n=0}^{N-1} x_1(n) e^{-j \frac{2\pi k n}{N}} \right] \left[ \sum_{l=0}^{N-1} x_2(l) e^{-j \frac{2\pi k l}{N}} \right] e^{j \frac{2\pi k m}{N}} \quad \dots(8)$$

Rearranging the summations and terms in Equation (8) we get,

$$y(m) = \frac{1}{N} \sum_{n=0}^{N-1} x_1(n) \sum_{l=0}^{N-1} x_2(l) \left[ \sum_{k=0}^{N-1} e^{-j \frac{2\pi k l}{N}} \cdot e^{-j \frac{2\pi k n}{N}} \cdot e^{j \frac{2\pi k m}{N}} \right]$$

$$\therefore y(m) = \frac{1}{N} \sum_{n=0}^{N-1} x_1(n) \sum_{l=0}^{N-1} x_2(l) \left[ \sum_{k=0}^{N-1} e^{+j \frac{2\pi k (m-n-l)}{N}} \right] \quad \dots(9)$$

Consider the last term of Equation (9); it can be written as,

$$e^{j \frac{2\pi k (m-n-l)}{N}} = \left[ e^{j \frac{2\pi (m-n-l)}{N}} \right]^k \quad \dots(10)$$

Now use the standard summation formula,

$$\sum_{k=0}^{N-1} a^k = \begin{cases} N & \text{for } a = 1 \\ \frac{1-a^N}{1-a} & \text{for } a \neq 1 \end{cases} \quad \dots(11)$$

$$\text{Let here, } a = e^{+j \frac{2\pi (m-n-l)}{N}} \quad \dots(12)$$

Now according to Equation (11) we will consider two cases :

### Case (i) : When $a = 1$

If  $(m-n-l)$  is multiple of  $N$  which means,

$$(m-n-l) = N, 2N, 3N, \dots \text{ then Equation (12) becomes,}$$

$$a = e^{+j 2\pi} = e^{+j 2\pi(2)} = e^{+j 2\pi(3)} \dots = 1$$

Thus when  $(m-n-l)$  is multiple of  $N$  (that means  $a = 1$ ) then according to Equation (11);

the third summation in Equation (9) becomes equal to N.

**Case (ii) : When  $a \neq 1$  :**

If  $a \neq 1$  that means if  $m - n - l$  is not multiple of N then according to Equation (11),

$$\sum_{k=0}^{N-1} a^k = \frac{1 - a^N}{1 - a} \quad \dots(13)$$

Putting Equation (12) in Equation (13) we get,

$$\sum_{k=0}^{N-1} \left( e^{+j2\pi(m-n-l)} \right)^k = \frac{1 - e^{+j2\pi(m-n-l)}}{1 - e^{+j2\pi(m-n-l)/N}} \quad \dots(14)$$

Here  $m$ ,  $n$  and  $l$  are integers; thus  $e^{+j2\pi(m-n-l)} = 1$  always. Thus R.H.S. of Equation (14) becomes zero when  $a \neq 1$ . So to get the result of Equation (11) we have to consider the condition  $a = 1$ . That means when  $m - n - l$  is multiple of N. For this condition, we have the result of summation equals to N. Thus Equation (9) becomes,

$$y(m) = \frac{1}{N} \sum_{n=0}^{N-1} x_1(n) \sum_{l=0}^{N-1} x_2(l) \cdot N$$

$$\therefore y(m) = \sum_{n=0}^{N-1} x_1(n) \sum_{l=0}^{N-1} x_2(l) \quad \dots(15)$$

We have obtained Equation (15) for the condition;  $(m - n - l)$  is multiple of N. This condition can be expressed as,

$$m - n - l = -pN \quad \dots(16)$$

Here 'p' is an integer and an integer can be positive or negative. For the simplicity we have considered negative integer. Now from Equation (16) we get,

$$l = m - n + pN \quad \dots(17)$$

Putting this value in Equation (15) we get,

$$y(m) = \sum_{n=0}^{N-1} x_1(n) \cdot x_2(m - n + pN) \quad \dots(18)$$

Here we have not considered the second summation of Equation (15). Because this summation is in terms of 'l' and exponential term is absent in Equation (18).

Now the term  $x_2(m - n + pN)$  indicates a periodic sequence with period N; this is because 'p' is an integer. This term also indicates that the periodic sequence is delayed by 'n' samples. We have studied that, if a sequence is periodic and delayed then it can be expressed as,

$$x_2(m - n + pN) = x_2((m - n))_N \quad \dots(19)$$

Here the R.H.S. term indicates circular shifting of  $x_2(n)$ . Putting this value in Equation (18).

we get,

$$y(m) = \sum_{n=0}^{N-1} x_1(n) \cdot x_2((m-n))_N \quad m = 0, 1, \dots, N-1 \quad \dots(20)$$

This is the equation of circular convolution. Since in this equation, the sequence  $x_2(n)$  is shifted circularly; this type of convolution is called as Circular Convolution.