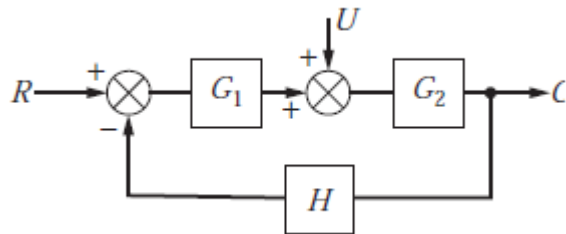


Root Locus

Chapter 14

CONCEPT OF ROOT LOCUS

- In this chapter, we develop a graphical method for finding the actual values of the roots of the characteristic equation, from which we can obtain the transient response of the system to an arbitrary forcing function.
- Illustrate the concept of a root locus diagram by considering the example presented



$$G_1 = K_c$$

$$G_2 = \frac{1}{(\tau_1 s + 1)(\tau_2 s + 1)}$$

$$H = \frac{1}{\tau_3 s + 1}$$

$$G = \frac{K_c}{(\tau_1 s + 1)(\tau_2 s + 1)(\tau_3 s + 1)}$$

$$G(s) = \frac{K}{(s - p_1)(s - p_2)(s - p_3)}$$

where $K = \frac{K_c}{\tau_1 \tau_2 \tau_3}$

$$p_1 = -\frac{1}{\tau_1} \quad p_2 = -\frac{1}{\tau_2} \quad p_3 = -\frac{1}{\tau_3}$$

- The terms $p_1, p_2, \text{ and } p_3$ are called the *poles* of the open-loop transfer function.
- A *pole* of $G(s)$ is any value of s for which $G(s)$ approaches infinity.
- Hence $p_1 = 1/\tau_1$ is a pole of $G(s)$.

The characteristic equation for the *closed-loop* system is

$$1 + \frac{K}{(s - p_1)(s - p_2)(s - p_3)} = 0$$

$$(s - p_1)(s - p_2)(s - p_3) + K = 0$$

Let by taking the value of $p_1 = -1, p_2 = -2, p_3 = -3$
and corresponding $K = 6K_c$

$$(s + 1)(s + 2)(s + 3) + K = 0$$

$$s^3 + 6s^2 + 11s + K + 6 = 0$$

which is third-order

- For any particular value of controller gain K_c , we can obtain the roots of the characteristic equation
- For example, if $K_c = 4.41$ ($K = 26.5$),

$$s^3 + 6s^2 + 11s + 32.5 = 0$$

$$r_1 = -5.10$$

$$r_2 = -0.45 - j2.5$$

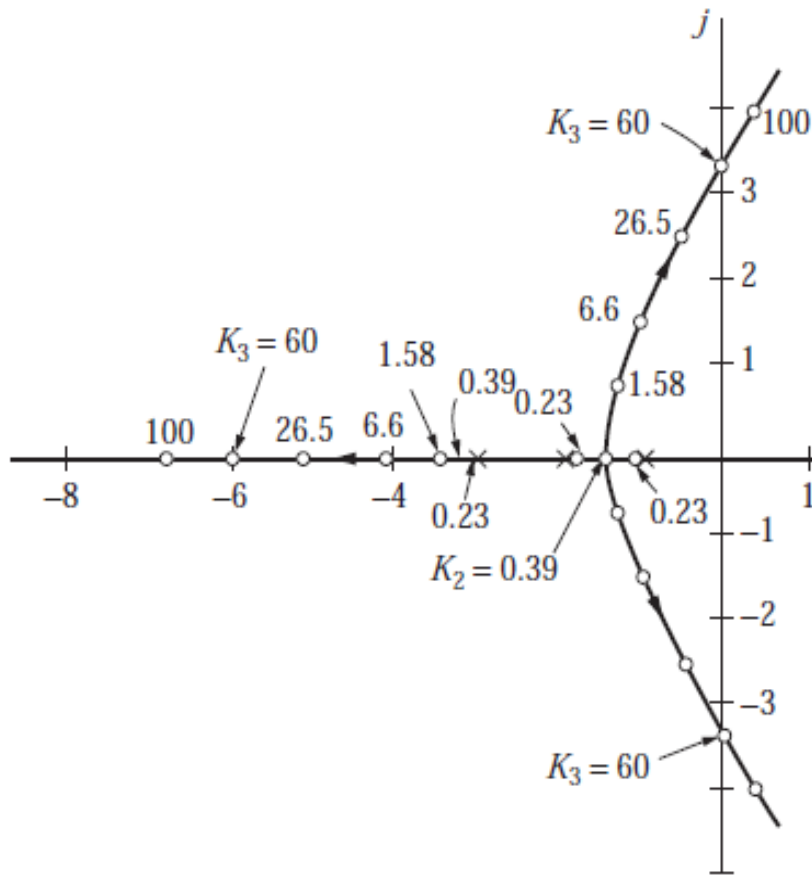
$$r_3 = -0.45 + j2.5$$

By selecting other values of K , other sets of roots are obtained

Roots of the characteristic equation $(s + 1)(s + 2)(s + 3) + K = 0$

$K = 6 K_c$	r_1	r_2	r_3
0	-3	-2	-1
0.23	-3.10	-1.75	-1.15
0.39	-3.16	-1.42	-1.42
1.58	-3.45	$-1.28 - j0.75$	$-1.28 + j0.75$
6.6	-4.11	$-0.95 - j1.5$	$-0.95 + j1.5$
26.5	-5.10	$-0.45 - j2.5$	$-0.45 + j2.5$
60.0	-6.00	$0.0 - j3.32$	$0.0 + j3.32$
100.0	-6.72	$0.35 - j4$	$0.35 + j4$

- For convenience, we may plot the roots r_1 , r_2 , and r_3 on the complex plane as K changes continuously. Such a plot is called a **root locus diagram**



Notice that there are three loci or *branches* corresponding to the three roots and that they “emerge” or begin (for $K=0$) at the poles of the open-loop transfer function ($-1, -2, -3$).

The direction of increasing K is indicated on the diagram by an arrow.

Also the values of K are marked on each locus. The root locus diagram for this system and others to follow is **symmetric with respect to the real axis**, and only the portion of the diagram in the upper half-plane need be drawn.

- The diagram reveals two critical values of K ; one is at K_2 ($K=0.39$) where two of the roots become equal,
- and the other is at K_3 ($K=60$) where two of the roots are pure imaginary.
- Thus, if the roots are all real, which occurs for $K < 0.39$, the response will be non-oscillatory.
- If two of the roots are complex and have negative real parts ($K_2 < K < K_3$), the response will include damped sinusoidal terms, which will produce an oscillatory response.
- If we adjust the controller gain such that $K > K_3$, two of the roots are complex and have positive real parts, and the response is a growing sinusoid.

