Angle Modulation

- ► We can encode information as either
 - Time varying phase
 - ► Time varying frequency
- ► Both result in angle modulation
- Both are very closely related
- Easy to do with an sdr, just another complex envelope!

Instantaneous Frequency

- In general, the frequency of a signal at an instant in time depends on the entire signal.
- For generalized sinusoids, we can use a simpler approach. Suppose

$$\varphi(t) = A\cos\theta(t).$$

Then $\theta(t)$ is the generalized angle. For a true sinusoid,

$$\theta(t) = \omega_c t + \theta_0 \,,$$

linear with slope ω_c and offset θ_0

- The generalized angle is *not* limited to $[0, 2\pi]$. Wrapping introduces discontinuities.
- Phase unwrapping is easy at an IF, where the phase changes are small from on sample to the next.

Instantaneous Frequency (cont.)

Instananeous frequency is derivative of generalized angle:

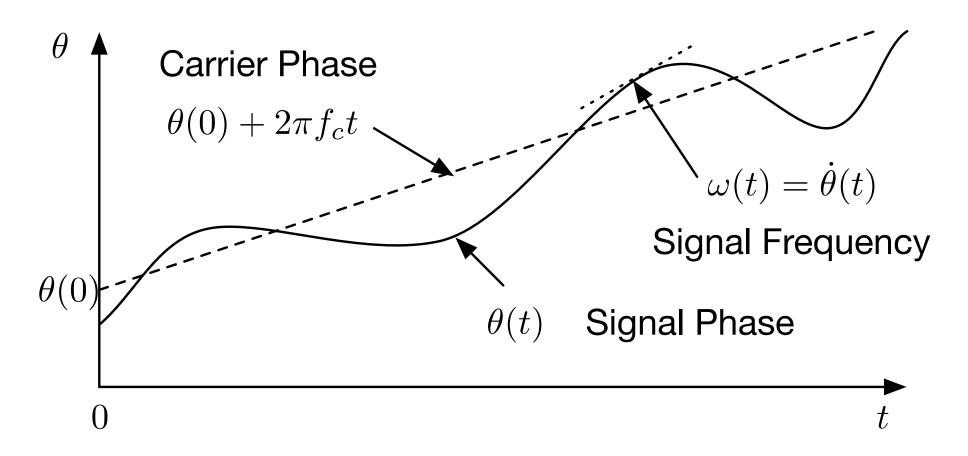
$$\omega_i(t) = \frac{d\theta}{dt} = \theta'(t)$$

► The phase is just the integral of the frequency

$$\theta(t) = \int_{-\infty}^{t} \omega_i(u) \, du = \theta(0) + \int_{0}^{t} \omega_i(u) \, du$$

- We can modulate a generalized sinusoid by using a signal m(t) to vary either $\theta(t)$ or $\omega_i(t)$.
- In either case, the frequency of the modulated signal changes as a function of m(t).

Instantaneous Frequency (cont.)



- Dashed line is the carrier phase
- Solid line is the phase of the transmitted signal
- Slope of the solid line is the instantaneous frequency of the transmitted signal.

Phase Modulation (PM)

▶ In PM, *phase* is varied *linearly* with m(t):

$$\theta(t) = 2\pi f_c t + k_p m(t)$$

which produces a transmitted signal

$$\varphi_{\rm PM}(t) = \cos(2\pi f_c t + k_p m(t))$$

The instantaneous frequency is

$$\omega_i(t) = \frac{d\theta}{dt} = 2\pi f_c + k_p \dot{m}(t)$$

- ▶ If m(t) varies rapidly, then the frequency deviations are larger.
- lacktriangle The bandwidth of the signal is determined by $\dot{m}(t)$, similar to AM.

Frequency Modulation (FM)

▶ In FM, frequency is is varied linear in m(t):

$$\omega_i(t) = 2\pi f_c + k_f m(t)$$

which produces a signal

$$\varphi_{\text{FM}}(t) = \cos\left(\left(2\pi f_c + k_f m(t)\right)t\right)\right)$$

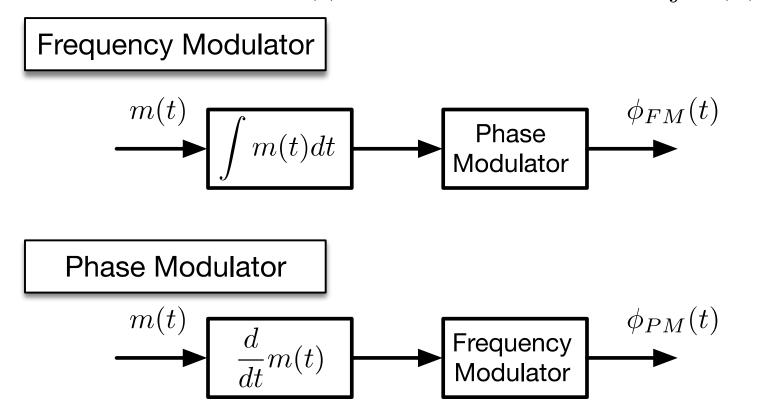
- The bandwidth of the signal is determined by the *amplitude* of $k_f m(t)$ (obvious), and also the bandwidth of m(t) (less obvious).
- The angle is

$$\theta(t) = \int_{-\infty}^{t} \left(2\pi f_c + k_f m(t)\right) du = 2\pi f_c t + k_f \int_{-\infty}^{t} m(u) du$$

We could apply this $\theta(t)$ to a phase modulator, and get exactly the same effect as applying $\omega_i(t)$ to a frequency modulator.

Relationship Between FM and PM

- ▶ Phase modulation of m(t) = frequency modulation of $\dot{m}(t)$.
- Frequency modulation of m(t) = phase modulation of $\int m(u) du$.



- We can produce both types of modulation with either modulator.
- Direct digital synthesis (DDS) chips will do this for you.

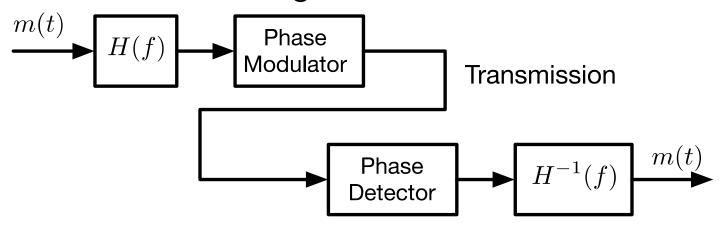
Generalized Angle Modulation

 \blacktriangleright We can generalize modulation by convolving the message signal with an impulse response h(t)

$$\varphi_{\rm EM}(t) = A\cos\left(2\pi f_c t + h(t) * m(t)\right)$$

This is a filter with a transfer function H(f).

We recover m(t) from phase by using inverse filter $H^{-1}(f)$. E.g., for FM the inverse of integration is differention.



- ightharpoonup PM $(h(t)=k_p\delta(t))$ and FM $(h(t)=k_fu(t))$ are special cases.
- Also used for pre-emphasis to improve noise characteristics, and pulse shaping to reduce signal bandwidth.